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APPLICATION NO.	FILING DATE	FIRST NAMED INVENTOR	ATTORNEY DOCKET NO.	CONFIRMATION NO.
10/770,395	02/04/2004	Ari Hottinen	60091.00269	2744
.32294	7590	04/20/2007	EXAMINER	
SQUIRE, SANDERS & DEMPSEY L.L.P.			HUANG, DAVID S	
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TYSONS CORNER, VA 22182			2609	
SHORTENED STATUTORY PERIOD OF RESPONSE		MAIL DATE	DELIVERY MODE	
3 MONTHS		04/20/2007	PAPER	

Please find below and/or attached an Office communication concerning this application or proceeding.

If NO period for reply is specified above, the maximum statutory period will apply and will expire 6 MONTHS from the mailing date of this communication.

Office Action Summary	Application No.	Applicant(s)	
	10/770,395	HOTTINEN ET AL.	
	Examiner	Art Unit	
	David Huang	2609	

-- The MAILING DATE of this communication appears on the cover sheet with the correspondence address --

Period for Reply

A SHORTENED STATUTORY PERIOD FOR REPLY IS SET TO EXPIRE 3 MONTH(S) OR THIRTY (30) DAYS, WHICHEVER IS LONGER, FROM THE MAILING DATE OF THIS COMMUNICATION.

- Extensions of time may be available under the provisions of 37 CFR 1.136(a). In no event, however, may a reply be timely filed after SIX (6) MONTHS from the mailing date of this communication.
- If NO period for reply is specified above, the maximum statutory period will apply and will expire SIX (6) MONTHS from the mailing date of this communication.
- Failure to reply within the set or extended period for reply will, by statute, cause the application to become ABANDONED (35 U.S.C. § 133). Any reply received by the Office later than three months after the mailing date of this communication, even if timely filed, may reduce any earned patent term adjustment. See 37 CFR 1.704(b).

Status

- 1) Responsive to communication(s) filed on 04 February 2004.
- 2a) This action is FINAL. 2b) This action is non-final.
- 3) Since this application is in condition for allowance except for formal matters, prosecution as to the merits is closed in accordance with the practice under *Ex parte Quayle*, 1935 C.D. 11, 453 O.G. 213.

Disposition of Claims

- 4) Claim(s) 1-12 is/are pending in the application.
- 4a) Of the above claim(s) _____ is/are withdrawn from consideration.
- 5) Claim(s) _____ is/are allowed.
- 6) Claim(s) 1-12 is/are rejected.
- 7) Claim(s) _____ is/are objected to.
- 8) Claim(s) _____ are subject to restriction and/or election requirement.

Application Papers

- 9) The specification is objected to by the Examiner.
- 10) The drawing(s) filed on 04 February 2004 is/are: a) accepted or b) objected to by the Examiner.
Applicant may not request that any objection to the drawing(s) be held in abeyance. See 37 CFR 1.85(a).
Replacement drawing sheet(s) including the correction is required if the drawing(s) is objected to. See 37 CFR 1.121(d).
- 11) The oath or declaration is objected to by the Examiner. Note the attached Office Action or form PTO-152.

Priority under 35 U.S.C. § 119

- 12) Acknowledgment is made of a claim for foreign priority under 35 U.S.C. § 119(a)-(d) or (f).
 - a) All b) Some * c) None of:
 1. Certified copies of the priority documents have been received.
 2. Certified copies of the priority documents have been received in Application No. _____.
 3. Copies of the certified copies of the priority documents have been received in this National Stage application from the International Bureau (PCT Rule 17.2(a)).

* See the attached detailed Office action for a list of the certified copies not received.

Attachment(s)

- 1) Notice of References Cited (PTO-892)
- 2) Notice of Draftsperson's Patent Drawing Review (PTO-948)
- 3) Information Disclosure Statement(s) (PTO/SB/08)
Paper No(s)/Mail Date 4/30/2004 and 4/6/2005
- 4) Interview Summary (PTO-413)
Paper No(s)/Mail Date. _____
- 5) Notice of Informal Patent Application
- 6) Other: _____

DETAILED ACTION

Priority

1. Receipt is acknowledged of papers submitted under 35 U.S.C. 119(a)-(d), which papers have been placed of record in the file.

Information Disclosure Statement

2. The references listed in the Information Disclosure Statements filed on April 30, 2004 and April 6, 2005 have been considered by the examiner (see attached PTO-1449 form or PTO/SB/08A and 08B forms).

Specification

3. The abstract of the disclosure is objected to because of the use of legal phraseology ("is disclosed"). Correction is required. See MPEP § 608.01(b).

Claim Objections

4. **Claims 4 and 5** are objected to because of the following informalities: the recitation of "the channel symbols transmitted using different transmit paths and different times" appears to be a new limitation lacking antecedent basis since it further specifies the channel symbols previously mentioned in claim 1. The language of the claims should be clarified to better convey the intended meaning. For examination on the merits, the claims will be interpreted as best understood. Appropriate correction is required.

5. **Claim 6** is objected to for being dependent on *claim 5*, and therefore contains the same problem explained above.

Claim Rejections - 35 USC § 103

6. The following is a quotation of 35 U.S.C. 103(a) which forms the basis for all obviousness rejections set forth in this Office action:

(a) A patent may not be obtained though the invention is not identically disclosed or described as set forth in section 102 of this title, if the differences between the subject matter sought to be patented and the prior art are such that the subject matter as a whole would have been obvious at the time the invention was made to a person having ordinary skill in the art to which said subject matter pertains. Patentability shall not be negated by the manner in which the invention was made.

7. **Claims 1, 7-8, 11 and 12** are rejected under 35 U.S.C. 103(a) as being unpatentable over Walton et al. (US Patent Application Publication 2002/0154705) in view of Tirkkonen et al. (PCT Application Publication WO 03/001728).

Regarding **claim 1**, Walton et al. discloses a transmission method comprising:
constructing layered channel symbols as linear combinations (In MIMO processing mode, each modulation symbol in the sub-channel represents a linear combination of modulation symbols, page 9, [0101]) of complex modulation symbols (bits grouped into modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM, page 9, [0101], where it is understood that PSK and QAM symbols are complex modulation symbols); and

transmitting the channel symbols via at least two transmit paths (RF modulated signals from modulators 114a through 114t are then transmitted from respective antennas 116a through 116t over communication links 118, page 3, [0032], Figure 1);

using, for at least one modulation symbol, a first non-zero total power for transmission on a first transmit path of the at least two transmit paths, and a second non-zero total power for transmission on a second transmit path of the at least two transmit paths, wherein the first and second total powers are not equal (when operating in diversity communications mode, if the pass

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loss from a particular antenna is great, transmission from this antenna can be reduced. Similarly, if transmission occurs over multiple sub-channels, less power may be transmitted on the sub-channel(s) experience the most path loss, page 13, [0137]).

However, Walton et al. fail to expressly disclose wherein using, when constructing the channel symbols, at least a first non-zero coefficient and a second non-zero coefficient in at least one layer when performing a linear combination, wherein the ratio of the first coefficient and the second coefficient is not a real number.

Tirkkonen et al. disclose the modulations symbols are subjected to complex diversity transform, wherein the symbol sequences are replaced, using complex diversity transform with super symbol sequences (page 16, lines 18-25). In an example, Tirkkonen et al. disclose the complex diversity transform is preferable implemented by s_3, s_4 being a unitary linear combination of QPSK symbols \hat{s}_3, \hat{s}_4 . The channel symbol $s_1 + s_3 = s_1 + \mu\hat{s}_3 + \nu\hat{s}_4$ is transmitted from the first antenna and the optimum values for coefficients μ, ν are complex values (page 13, lines 9-29). Therefore, the ratio between the optimum μ and ν is not a real number (imaginary).

Therefore it would have been obvious to one of ordinary skill in the art at the time the invention was made to modify Walton et al. with the complex diversity transform taught by Tirkkonen et al. since the super symbol sequences generated by the complex diversity transform endure fading and interference on the transmission path and decrease the effect of inter-symbol sequence interference on the detection of the symbol sequences (page 16, lines 24-28).

Regarding **claim 7**, Walton et al. disclose a transmitter comprising:

antenna means for achieving two transmit paths for transmission of a signal (RF modulated signals from modulators 114a through 114t are then transmitted from respective antennas 116a through 116t over communications links 118 to system 120, page 3, [0032], and Figure 1);

means for modulating (Data processor 320 thus receives and processes the encoded data corresponding to K channel data streams to provide NT modulation symbol vectors, V_1 through V_{NT} , one modulation symbol vector for each transmit antenna, page 9, [0104], Figure 3) the signal to be transmitted into complex modulation symbols (bits grouped into modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM, page 9, [0101], where it is understood that PSK and QAM symbols are complex);

means for constructing layered channel symbols (Data processor 320 thus receives and processes the encoded data corresponding to K channel data streams to provide NT modulation symbol vectors, V_1 through V_{NT} , one modulation symbol vector for each transmit antenna, page 9, [0104], Figure 3) as linear combinations of the complex modulation symbols (In the MIMO processing mode, each modulation symbol in the sub-channel represents a linear combination of modulation symbols, page 9, [0101]); and

means for (power control can be achieved with a feedback mechanism similar to that used in the CDMA system, page 13, [0138]) transmitting the channel symbols by using, for at least one modulation symbol, a first non-zero total power for transmission on a first transmit path, and a second non-zero total power for transmission on a second transmit path, wherein the first and second total powers are not equal (when operating in diversity communications mode, if the pass loss from a particular antenna is great, transmission from this antenna can be reduced.

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Similarly, if transmission occurs over multiple sub-channels, less power may be transmitted on the sub-channel(s) experience the most path loss, page 13, [0137]).

However Walton et al. fail to expressly disclose means for constructing channel symbols by using at least a first non-zero coefficient and a second non-zero coefficient in at least one layer when performing the linear combinations, wherein the ratio of the first and second coefficient is not a real number.

Tirkkonen et al. disclose the modulations symbols are first applied to transform means 304 in which the symbols are subjected to complex diversity transform, wherein the symbol sequences are replaced, using complex diversity transform with super symbol sequences (page 16, lines 18-25). In an example, Tirkkonen et al. disclose the complex diversity transform is preferable implemented by s_3, s_4 being a unitary linear combination of QPSK symbols \hat{s}_3, \hat{s}_4 . The channel symbol $s_1 + s_3 = s_1 + \mu\hat{s}_3 + \nu\hat{s}_4$ is transmitted from the first antenna and the optimum values for coefficients μ, ν are complex values (page 13, lines 9-29). Therefore, the ratio between the optimum μ and ν is not a real number (imaginary).

Therefore, it would have been obvious to one of ordinary skill in the art at the time the invention was made to provide Walton et al. with the transform means 304 taught by Tirkkonen et al. since the super symbol sequences generated by the complex diversity transform endure fading and interference on the transmission path and decrease the effect of inter-symbol sequence interference on the detection of the symbol sequences (page 16, lines 24-28).

Regarding **claims 8, 11 and 12**, Walton et al. disclose a transmitter, implemented in either a base station or terminal equipment of a cellular radio system (page 3, [0036], Figure 1), comprising:

an antenna system for achieving two transmit paths for transmission of a signal (RF modulated signals from modulators 114a through 114t are then transmitted from respective antennas 116a through 116t over communications links 118 to system 120, page 3, [0032], and Figure 1);

a first modulator for modulating (Data processor 320 thus receives and processes the encoded data corresponding to K channel data streams to provide NT modulation symbol vectors, V_1 through V_{NT} , one modulation symbol vector for each transmit antenna, page 9, [0104], Figure 3) the signal to be transmitted into complex modulation symbols (bits grouped into modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM, page 9, [0101], where it is understood that PSK and QAM symbols are complex);

a second modulator for constructing layered channel symbols (Data processor 320 thus receives and processes the encoded data corresponding to K channel data streams to provide NT modulation symbol vectors, V_1 through V_{NT} , one modulation symbol vector for each transmit antenna, page 9, [0104], Figure 3) as linear combinations of the complex modulation symbols (In the MIMO processing mode, each modulation symbol in the sub-channel represents a linear combination of modulation symbols, page 9, [0101]); and

the second modulator (Data processor 320, figure 3) and the antenna system (116a – 116t, Figure 1) are configured to transmit the channel symbols by using, for at least one modulation symbol, a first non-zero total power for transmission on a first transmit path, and a second non-zero total power for transmission on a second transmit path, wherein the first and second total powers are not equal (power control can be performed on each channel data stream, on each sub-channel, and on each antenna. When operating in the diversity communications mode, if the path

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loss from a particular antenna is great, transmission from this antenna can be reduced since little may be gained at the receiver unit. Similarly, if transmission occurs over multiple sub-channels, less power may be transmitted on the sub-channel(s) experiencing the most path loss, page 13, [0137]).

However, Walton et al. fail to expressly disclose wherein the second modulator is configured to construct the channel symbols by using at least a first non-zero coefficient and a second non-zero coefficient in at least one layer when performing the linear combination, wherein the ratio of the first and second coefficient is not a real number.

Tirkkonen et al. disclose the modulations symbols are first applied to transform means 304 in which the symbols are subjected to complex diversity transform, wherein the symbol sequences are replaced, using complex diversity transform with super symbol sequences (page 16, lines 18-25). In an example, Tirkkonen et al. disclose the complex diversity transform is preferable implemented by s_3, s_4 being a unitary linear combination of QPSK symbols \hat{s}_3, \hat{s}_4 . The channel symbol $s_1 + s_3 = s_1 + \mu\hat{s}_3 + \nu\hat{s}_4$ is transmitted from the first antenna and the optimum values for coefficients μ, ν are complex values (page 13, lines 9-29). Therefore, the ratio between the optimum μ and ν is not a real number (imaginary).

Therefore, it would have been obvious to one of ordinary skill in the art at the time the invention was made to provide Walton et al. with the transform means 304 taught by Tirkkonen et al. since the super symbol sequences generated by the complex diversity transform endure fading and interference on the transmission path and decrease the effect of inter-symbol sequence interference on the detection of the symbol sequences (page 16, lines 24-28).

8. **Claims 2 and 9** are rejected under 35 U.S.C. 103(a) as being unpatentable over Walton et al. (US Patent Application Publication 2002/0154705) in view of Tirkkonen et al. (PCT Application Publication WO 03/001728) as applied to *claims 1 and 7* above, and further in view of Sampath (US Patent Application Publication 2003/0043929).

Regarding **claims 2 and 9**, the combination of Walton et al. and Tirkkonen et al. discloses everything claimed as applied above (see *claims 1 and 7*), but fails to expressly disclose the step and associated transmitter means for using at least one complex precoder matrix that comprises at least two non-zero elements that have different transmission powers.

Sampath teaches a preprocessor can also scale input symbol streams, depending upon system implementation issues that will be explained below. For example, in the case of scaling matrix dependence upon a BER/SR requirement at the receiver, if a first symbol stream requires a lower BER than a second symbol stream, then the preprocessor must allocate more power to the first stream (that is, scales it higher) and less power to the second stream (that is, scales it lower). If a first symbol stream includes a higher transmission order (higher order QAM) than a second symbol stream, the preprocessor can allocate more power to the first stream and less power to the second stream (page 5, [0066]-[0069]).

Therefore it would have been obvious to one of ordinary skill in the art at the time the invention was made to provide the combination of Walton et al. and Tirkkonen et al. with the preprocessor taught by Sampath since it would allocate more power to symbol streams including higher transmission orders thereby improving symbol reception.

9. **Claims 3 and 10** rejected under 35 U.S.C. 103(a) as being unpatentable over Walton et al. (US Patent Application Publication 2002/0154705) in view of Tirkkonen et al. (PCT

Application Publication WO 03/001728) as applied to *claims 1 and 7* above, and further in view of Lott et al. (US Patent Application Publication 2004/0120287).

Regarding **claims 3 and 10**, the combination of Walton et al. and Tirkkonen et al. discloses everything claimed as applied above (see *claims 1 and 7*), but fails to expressly disclose the step and associated transmitter means for using at least one real precoder matrix, wherein a transmission power ratio between symbols transmitted at different times within a layer is at least 2/8.

However, Walton et al. disclose that in MIMO processing mode, each modulation symbol in the sub-channel represents a linear combination of modulation symbols and bits are grouped into modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM, page 9, [0101], where it is common knowledge that BPSK symbols carry 1-bit of information, while 16-QAM symbols carry 4-bits of information.

Sampath teaches a preprocessor can also scale input symbol streams, depending upon system implementation issues that will be explained below. For example, in the case of scaling matrix dependence upon a BER/SR requirement at the receiver, if a first symbol stream requires a lower BER than a second symbol stream, then the preprocessor must allocate more power to the first stream (that is, scales it higher) and less power to the second stream (that is, scales it lower). If a first symbol stream includes a higher transmission order (higher order QAM) than a second symbol stream, the preprocessor can allocate more power to the first stream and less power to the second stream (page 5, [0066]-[0069]).

It is also known in the art that since power required to transmit data is roughly proportional to the data rate, increasing the data rate would also increase transmission power, as evidenced by Lott et al. (page 4, [0042]).

Therefore it would have been obvious to one of ordinary skill in the art at the time the invention was made to provide the combination of Walton et al. and Tirkkonen et al. with the preprocessor taught by Sampath since it would adaptively allocate more power to symbol streams including symbols with higher order modulations, thereby improving symbol reception. It would also have been obvious to one of ordinary skill in the art at the time was made to specify the power ratio to be 2/8 or ¼ since sub-channel modulation symbols are a linear combination of complex modulation symbols including BPSK and 16-QAM which carry 1 and 4 bits of information, respectively, and the transmission power is proportional to the data rate.

10. **Claim 4-6** is rejected under 35 U.S.C. 103(a) as being unpatentable over Walton et al. (US Patent Application Publication 2002/0154705) in view of Tirkkonen et al. (PCT Application Publication WO 03/001728) as applied to claim 1 above, and further in view of Brailean et al (US Patent 6,002,715).

Regarding **claim 4**, Walton et al. disclose everything claimed as applied above (see *claim 1*), but fail to expressly disclose wherein the channel symbols transmitted using different transmit paths and different times form equidistant QAM constellations.

Nevertheless, Walton et al. teach symbols in the sub-channel represent linear combinations of modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM (page 9, [0101]). Thus, it is implicit in Walton et al.'s disclosure that the sub-channel symbols consist of linear combinations of QAM modulation symbols.

It is also well known in the art that 16 QAM constellation can be arranged in three groups such that, within any group, each symbol is equidistant from the origin, as is evidenced by Brailean et al. (column 2, line 63 – column 3, line 11; see table 1). In other words, the 16 QAM constellation consists of three groups of equidistant QAM constellations.

Therefore it would have been obvious to one of ordinary skill in the art, at the time the invention was made to modify Walton et al. so that the transmitted sub-channel symbols form equidistant QAM constellations because the use of 16 QAM as the specified modulation scheme is an engineering expedient suggested by Walton et al., and also because 16 QAM is well known in the art to consist of 3 groups of equidistant QAM constellations.

Regarding **claim 5**, Walton et al. disclose everything claimed as applied above (see *claim 1*), but fail to expressly disclose wherein the channel symbols transmitted using different transmit paths and different times form a lattice.

Nevertheless, Walton et al. teach symbols in the sub-channel represent linear combinations of modulation symbols using a particular modulation scheme e.g., M-PSK or M-QAM (page 9, [0101]). Thus, it is implicit in Walton et al.’s disclosure that the sub-channel symbols consist of linear combinations of QAM modulation symbols.

It is also well known in the art that 4-QAM forms a constellation in which each symbol is equidistant from the origin, as is evidenced by Brailean et al. (core symbols, column 3, lines 34-37; see table 1).

Further, “lattice” is defined as a regular geometrical arrangement of points or objects over an area or in space (Merriam-Webster’s Medical Dictionary, 2002).

Therefore it would have been obvious to one of ordinary skill in the art, at the time the invention was made to modify Walton et al. so that the transmitted sub-channel symbols form a lattice because the use of QAM as the specified modulation scheme is an engineering expedient suggested by Walton et al., and also because 4-QAM forms an equidistant constellation, and therefore a lattice.

Regarding **claim 6**, the combination of Walton et al. as evidenced by Brailean et al. disclose everything claimed as applied above (see *claim 5*), and further disclose wherein the lattice is equidistant (4-QAM constellation is equidistant).

Conclusion

Any inquiry concerning this communication or earlier communications from the examiner should be directed to David Huang whose telephone number is (571) 270-1798. The examiner can normally be reached on Monday - Friday, 8:00 a.m. - 5:00 p.m., EST.

If attempts to reach the examiner by telephone are unsuccessful, the examiner's supervisor, Eliseo Ramos-Feliciano can be reached on (571) 272-7925. The fax phone number for the organization where this application or proceeding is assigned is 571-273-8300.

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Information regarding the status of an application may be obtained from the Patent Application Information Retrieval (PAIR) system. Status information for published applications may be obtained from either Private PAIR or Public PAIR. Status information for unpublished applications is available through Private PAIR only. For more information about the PAIR system, see <http://pair-direct.uspto.gov>. Should you have questions on access to the Private PAIR system, contact the Electronic Business Center (EBC) at 866-217-9197 (toll-free). If you would like assistance from a USPTO Customer Service Representative or access to the automated information system, call 800-786-9199 (IN USA OR CANADA) or 571-272-1000.

DSH/dsh

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4-16-07

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